

Comparative Study between Two Important Nonlinear Methodologies for Continuum Robot Manipulator Control

Alireza Salehi, Farzin Piltan, Mahdi Mirshekaran, Meysam Kazeminasab, Zahra Esmaeili

Institute of Advance Science and Technology, Intelligent control and Robotics Lab. IRAN SSP, Shiraz, Iran *E-mail: Piltan_f@iranssp.com; http://WWW.IRANSSP.COM*

Abstract- This research focuses on the basic concepts of continuum robot manipulator and control methodology. OCTARM Continuum robot manipulator is a 6 DOF serial robot manipulator. From the control point of view, robot manipulator divides into two main parts i.e. kinematics and dynamic parts. The dynamic parameters of this system are highly nonlinear. To control of this system nonlinear control methodology (computed torque controller and sliding mode controller) is introduced. Computed torque controller (CTC) is an influential nonlinear controller to certain systems which it is based on feedback linearization and computes the required arm torques using the nonlinear feedback control law. When all dynamic and physical parameters are known computed torque controller works superbly; practically a large amount of systems have uncertainties and sliding mode controller reduce this kind of limitation. Sliding mode controller (SMC) is a significant nonlinear controller under condition of partly uncertain dynamic parameters of system. This controller is used to control of highly nonlinear systems especially for robot manipulators, because this controller is a robust and stable. Comparative study between computed torque controller and sliding mode controller is introduced in this research.

Index Terms- Continuum Robot Manipulator, Sliding Mode Control, Computed Torque Control, Nonlinear Methodology, Artificial Intelligence Technique

I. Introduction and Background

The international organization defines the robot as "an automatically controlled, reprogrammable, multipurpose manipulator with three or more axes." The institute of robotic in The United States Of America defines the robot as "a reprogrammable, multifunctional manipulator design to move material, parts, tools, or specialized devices through various programmed motions for the performance of variety of tasks" [1-20].

Robot manipulator is a collection of links that connect to each other by joints, these joints can be revolute and prismatic that revolute joint has rotary

motion around an axis and prismatic joint has linear motion around an axis. Each joint provides one or more degrees of freedom (DOF). From the mechanical point of view, robot manipulator is divided into two main groups, which called; serial robot links and parallel robot links. In serial robot manipulator, links and joints is serially connected between base and final frame (endeffector). Parallel robot manipulators have many legs with some links and joints, where in these robot manipulators base frame has connected to the final frame. Most of industrial robots are serial links, which in n degrees of freedom serial link robot manipulator the axis of the first three joints has a known as major axis, these axes show the position of end-effector, the axis number four to six are the minor axes that use to calculate the orientation of end-effector and the axis number seven to n use to reach the avoid the difficult conditions (e.g., surgical robot and space robot manipulator). Kinematics is an important subject to find the relationship between rigid bodies (e.g., position and orientation) and end-effector in robot manipulator. The mentioned topic is very important to describe the three areas in robot manipulator: practical application such as trajectory planning, essential prerequisite for some dynamic description such as Newton's equation for motion of point mass, and control purposed therefore kinematics play important role to design accurate controller for robot manipulators. Robot manipulator kinematics is divided into two main groups: forward kinematics and inverse kinematics where forward kinematics is used to calculate the position and orientation of end-effector with given joint parameters (e.g., joint angles and joint displacement) and the activated position and orientation of end-effector calculate the joint variables in Inverse Kinematics[6]. Dynamic modeling of robot manipulators is used to describe the behavior of robot manipulator such as linear or nonlinear dynamic behavior, design of model based controller such as pure sliding mode controller and pure computed torque controller which design these controllers are based on nonlinear dynamic equations, and for simulation. The dynamic modeling describes the relationship between joint motion, velocity, and accelerations to force/torque or current/voltage and also it can be used to describe the particular dynamic effects (e.g., inertia, coriolios, centrifugal, and the other

parameters) to behavior of system[21-33]. The continuum robot has a nonlinear and uncertain dynamic parameters serial link 6 degrees of freedom (DOF) robot manipulator. Design a nonlinear robust controller is major subject in this project.

Controller is a device which can sense information from linear or nonlinear system (e.g., robot manipulator) to improve the systems performance [34-55]. The main targets in designing control systems are stability, good disturbance rejection, and small tracking error[50-55]. Several industrial robot manipulators are controlled by linear methodologies (e.g., Proportional-Derivative (PD) controller, Proportional- Integral (PI) controller or Proportional- Integral-Derivative (PID) controller), but when robot manipulator works with various payloads and have uncertainty in dynamic models this technique has limitations. In some applications robot manipulators are used in an unknown and unstructured environment. therefore strong mathematical tools used in new control methodologies to design nonlinear robust controller with an acceptable performance (e.g., minimum error, good trajectory, disturbance rejection). Sliding mode controller and computed torque controller are influential nonlinear controllers to certain and uncertain systems which they work based on system's dynamic model. Sliding mode controller is a powerful nonlinear robust controller under condition of partly uncertain dynamic parameters of system [7-21]. This controller is used to control of highly nonlinear systems especially for robot manipulators. Chattering phenomenon and nonlinear equivalent dynamic formulation in uncertain dynamic parameter are two main drawbacks in pure sliding mode controller [20-32]. The chattering phenomenon problem in pure sliding mode controller is reduced by using linear saturation boundary layer function but prove the stability is very difficult. The nonlinear equivalent dynamic formulation problem in uncertain system is solved by using intelligent theorem [8-29]. Computed torque controller (CTC) is a powerful nonlinear controller which it widely used in control of robot manipulator. It is based on feedback linearization and computes the required arm torques using the nonlinear feedback control law. This controller works very well when all dynamic and physical parameters are known but when the robot manipulator has variation in dynamic parameters, in this situation the controller has no acceptable performance[14-28]. In practice, most of physical systems (e.g., robot manipulators) parameters are unknown or time variant, therefore, computed torque like controller used to compensate dynamic equation of robot manipulator[1- 6]. Research on computed torque controller is significantly growing on

robot manipulator application which has been reported in [1-16].

This paper is organized as follows; second part focuses on the modeling dynamic formulation based on Lagrange methodology. Third part is focused on the design sliding mode controller and computed torque controller which can be used to reduce the error, increase the performance quality and increase the robustness and stability. Simulation result and discussion is illustrated in forth part which based on trajectory following and disturbance rejection. The last part focuses on the conclusion and compare between this method and the other ones.

II. Theory

Dynamic Formulation of Continuum Robot: The Continuum section analytical model developed here consists of three modules stacked together in series. In general, the model will be a more precise replication of the behavior of a continuum arm with a greater of modules included in series. However, we will show that three modules effectively represent the dynamic behavior of the hardware, so more complex models are not motivated. Thus, the constant curvature bend exhibited by the section is incorporated inherently within the model. The model resulting from the application of Lagrange's equations of motion obtained for this system can be represented in the form

$$F_{coeff} \,\underline{\tau} = D\left(\underline{q}\right) \underline{\ddot{q}} + C\left(\underline{q}\right) \underline{\dot{q}} + G\left(\underline{q}\right) \tag{1}$$

where τ is a vector of input forces and q is a vector of generalized co-ordinates. The force coefficient matrix F_{coeff} transforms the input forces to the generalized forces and torques in the system. The inertia matrix, D is composed of four block matrices. The block matrices that correspond to pure linear accelerations and pure angular accelerations in the system (on the top left and on the bottom right) are symmetric. The matrix C contains coefficients of the first order derivatives of the generalized co-ordinates. Since the system is nonlinear, many elements of C contain first order derivatives of the generalized co-ordinates. The remaining terms in the dynamic equations resulting from gravitational potential energies and spring energies are collected in the matrix G. The coefficient matrices of the dynamic equations are given below,

$$Fcoeff = \begin{bmatrix} 1 & 1 & cos(\theta_1) & cos(\theta_1) & cos(\theta_1 + \theta_2) & cos(\theta_1 + \theta_2) \\ 0 & 0 & 1 & 1 & cos(\theta_2) & cos(\theta_2) \\ 0 & 0 & 0 & 0 & 1 & 1 \\ 1/2 & -1/2 & 1/2 & -1/2 & 1/2 + s_2 sin(\theta_2) & -1/2 + s_2 sin(\theta_2) \\ 0 & 0 & 1/2 & -1/2 & 1/2 & -1/2 \\ 0 & 0 & 0 & 0 & 1/2 & -1/2 \end{bmatrix}$$
(2)

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D	(q)					
	$\begin{array}{c} & m_1 + m_2 \\ & + m_3 \end{array}$	$\begin{array}{l} m_2 cos(\theta_1) \\ + m_3 cos(\theta_1) \end{array}$	$m_3 cos(\theta_1 + \theta_2)$	$\begin{array}{l} -m_2 s_2 sin(\theta_1) \\ -m_3 s_2 sin(\theta_1) \\ -m_3 s_3 sin(\theta_1 + \theta_2) \end{array}$	$-m_3s_3sin(\theta_1+\theta_2)$	0
	$\begin{array}{l} m_2 cos(\theta_1) \\ + m_3 cos(\theta_1) \end{array}$	$m_2 + m_3$	$m_3 cos(\theta_2)$	$-m_3s_3sin(\theta_2)$	$-m_3s_3sin(\theta_2)$	0
	$m_3 cos(\theta_1 + \theta_2)$	$m_3 cos(\theta_2)$	m_3	$m_3 s_3 sin(\theta_2)$	0	0
=	$\begin{array}{l} -m_2 s_2 sin(\theta_1) \\ -m_3 s_2 sin(\theta_1) \\ -m_3 s_3 sin(\theta_1 + \theta_2) \end{array}$	$-m_3s_3sin(\theta_2)$	$m_3 s_2 sin(\theta_2)$	$m_2 s_2^2 + I_1 + I_2 + I_3 + m_3 s_2^2 + m_3 s_3^2 + 2m_3 s_3 cos(\theta_2) s_2$	$I_2 + m_3 s_3^2 + I_3 + m_3 s_3 cos(\theta_2) s_2$	I ₃
	$-m_3s_3sin(\theta_1+\theta_2)$	$-m_3s_3sin(\theta_2)$	0	$I_2 + m_3 s_3^2 + I_3 + m_3 s_3 cos(\theta_2) s_2 I$	$I_2 + m_3 s_3^2 + I_3$	I ₃
	o	0	0	I ₃	I ₃	I_3

C(q) $-m_{2}s_{2}$ $cos(\theta_1)(\dot{\theta}_1)$ $+(1/2)(c_{11}+c_{21})$ $-2m_3sin(\theta_1+\theta_2)$ $-2m_2sin(\theta_1)\dot{\theta}_1$ $-m_3s_2$ $-m_3s_3sin(\theta_1+\theta_2)$ 0 $c_{11} + c_{21}$ $-2m_3sin(\theta_1)\dot{\theta}_1$ $(\dot{\theta}_1 + \dot{\theta}_2)$ $cos(\theta_1)(\dot{\theta}_1)$ $-m_{3}s_{3}$ $cos(\theta_1 + \theta_2)(\dot{\theta}_1)$ $-m_3s_3(\dot{\theta}_1)$ +(1/2) $-2m_{3}s_{3}$ $-2m_3sin(\theta_2)$ $cos(\theta_2)(\dot{\theta}_1)$ $(c_{12} + c_{22})$ 0 0 $c_{12} + c_{22}$ $-m_3s_2(\theta_1)$ $(\dot{\theta}_1 + \dot{\theta}_2)$ $-m_{3}s_{3}$ $-m_{3}s_{3}$ $cos(\theta_2)(\dot{\theta}_2)$ $cos(\theta_2)(\dot{\theta}_1)$ = $-m_3s_3s_2$ $-2m_3s_3(\dot{\theta}_1)$ (1/2) $cos(\theta_2)(\dot{\theta}_1)$ $2m_3sin(\theta_2)(\dot{\theta}_1)$ 0 $c_{13} + c_{23}$ $-m_3s_3(\dot{\theta}_2)$ $(c_{13} + c_{23})$ $-m_3s_3(\dot{\theta}_1)$ $2m_3s_3s_2$ $2m_3s_3cos(\theta_2)(\dot{\theta}_1)$ $2m_3s_3(\dot{\theta}_1+\dot{\theta}_2)$ $m_3 s_3 s_2$ (1/2) $sin(\theta_2)(\dot{\theta}_2)$ $-2m_3s_2(\dot{\theta}_1)$ $-2m_3s_2cos(\theta_2)$ 0 $sin(\theta_2)(\dot{\theta}_2)$ $+(1^2/4)$ $(c_{11} + c_{21})$ $(\dot{\theta}_1 + \dot{\theta}_2)$ $+2m_2s_2(\dot{\theta}_1)$ $(c_{11} + c_{21})$ $2m_{3}s_{3}$ $m_3 s_3 s_2$ $(1/2)(c_{12}+c_{22})+$ $(1^2/4)$ 0 0 $2m_3s_3cos(\theta_2)(\dot{\theta}_1)$ $\left(\dot{\theta}_1 + \dot{\theta}_2\right)$ $sin(\theta_2)(\dot{\theta}_1)$ $(c_{12} + c_{22})$ $(1^2/4)$ 0 0 $(1/2)(c_{13}-c_{23})$ 0 0 $(c_{13} + c_{23})$

$$G\left(\underline{q}\right) = \begin{bmatrix} -m_{1}g - m_{2}g + k_{11}(s_{1} + (1/2)\theta_{1} - s_{01}) + k_{21}(s_{1} - (1/2)\theta_{1} - s_{01}) - m_{3}g \\ -m_{2}g\cos(\theta_{1}) + k_{12}(s_{2} + (1/2)\theta_{2} - s_{02}) + k_{22}(s_{2} - (1/2)\theta_{2} - s_{02}) - m_{3}g\cos(\theta_{1}) \\ -m_{3}g\cos(\theta_{1} + \theta_{2}) + k_{13}(s_{3} + (1/2)\theta_{3} - s_{03}) + k_{23}(s_{3} - (1/2)\theta_{3} - s_{03}) \\ m_{2}s_{2}g\sin(\theta_{1}) + m_{3}s_{3}g\sin(\theta_{1} + \theta_{2}) + m_{3}s_{2}g\sin(\theta_{1}) + k_{11}(s_{1} + (1/2)\theta_{1} - s_{01})(1/2) \\ + k_{21}(s_{1} - (1/2)\theta_{1} - s_{01})(-1/2) \\ m_{3}s_{3}g\sin(\theta_{1} + \theta_{2}) + k_{12}(s_{2} + (1/2)\theta_{2} - s_{02})(1/2) + k_{22}(s_{2} - (1/2)\theta_{2} - s_{02})(-1/2) \\ k_{13}(s_{3} + (1/2)\theta_{3} - s_{03})(1/2) + k_{23}(s_{3} - (1/2)\theta_{3} - s_{03})(-1/2) \end{bmatrix}$$

$$(5)$$

(3)

(4)

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III. Methodology

SLIDING MODE CONTROLLER: One of the significant challenges in control algorithms is a linear behavior controller design for nonlinear systems. When system works with various parameters and hard nonlinearities this technique is very useful in order to be implemented easily but it has some limitations such as working near the system operating point[2]. Some of robot manipulators which work in industrial processes are controlled by linear PID controllers, but the design of linear controller for robot manipulators is extremely difficult because they are nonlinear, uncertain and MIMO[33-55]. To reduce above challenges the nonlinear robust controllers is used to systems control. One of the powerful nonlinear robust controllers is sliding mode controller (SMC), although this controller has been analyzed by many researchers but the first proposed was in the 1950 [7]. This controller is used in wide range areas such as in robotics, in control process, in aerospace applications and in power converters because it has an acceptable control performance and solve some main challenging topics in control such as resistivity to the external disturbance. the lyapunov formulation can be written as follows.

$$V = \frac{1}{2} S^T . D . S \tag{6}$$

The derivation of V can be determined as,

$$\dot{V} = \frac{1}{2} S^T \cdot \dot{D} \cdot S + S^T D \dot{S}$$
⁽⁷⁾

The dynamic equation of robot manipulator can be written based on the sliding surface as

$$D\dot{S} = -VS + D\dot{S} + VS + G - \tau \tag{8}$$

It is assumed that

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$$S^{T}(\dot{D}-2V)S=0 \tag{9}$$

by substituting (8) in (7)

$$\dot{V} = \frac{1}{2}S^T\dot{D}S - S^TVS + S^T(D\dot{S} + VS + G - \tau)$$

$$\tau) = S^T(D\dot{S} + VS + G - \tau)$$
(10)

Suppose the control input is written as follows

$$\hat{\tau} = \hat{\tau_{eq}} + \hat{\tau_{dis}} = \left[\widehat{D^{-1}} (\widehat{V} + \widehat{G}) + \dot{S} \right] \widehat{D} + (11)$$

K. sgn(S) + K_vS

by replacing the equation (11) in (10)

$$\dot{V} = S^{T}(D\dot{S} + VS + G - D\dot{S} - VS - G - K_{\nu}S - Ksgn(S) = S^{T}(D\dot{S} + VS + G - Ksgn(S))$$
(12)

It is obvious that

$$\begin{aligned} \left| \widetilde{D}\dot{S} + \widetilde{V}S + \widetilde{G} - K_{v}S \right| &\leq \left| \widetilde{D}\dot{S} \right| + \left| \widetilde{V}S \right| + \\ \left| \widetilde{G} \right| + \left| K_{v}S \right| \end{aligned} \tag{13}$$

The Lemma equation in robot manipulator system can be written as follows

$$K_{u} = \left[\left| \widetilde{DS} \right| + |VS| + |G| + |K_{v}S| + \eta \right]_{i},$$

 $i = 1, 2, 3, 4, ...$
(14)

The equation (14) can be written as

$$K_{u} \ge \left| \left[\widetilde{D}\dot{S} + VS + G - K_{v}S \right]_{i} \right| + \eta_{i}$$
⁽¹⁵⁾

Therefore, it can be shown that

$$\dot{V} \le -\sum_{i=1}^{n} \eta_i \left| S_i \right| \tag{16}$$

Based on above discussion, the control law for a multi degrees of freedom robot manipulator is written as:

$$U = U_{eq} + U_r \tag{17}$$

Where, the model-based component U_{eq} is the nominal dynamics of systems and U_{eq} can be calculate as follows:

$$\boldsymbol{U}_{eq} = \left[\boldsymbol{D}^{-1}(\boldsymbol{f} + \boldsymbol{C} + \boldsymbol{G}) + \dot{\boldsymbol{S}}\right]\boldsymbol{D}$$
(18)

and **U**_{SWITCH} is computed as;

$$U_{sat} = K \cdot \text{SGN}(S) \tag{19}$$

by replace the formulation (9) in (7) the control output can be written as;

$$U = U_{eq} + K. \, \mathrm{SGN}(S) \tag{20}$$

By (10) and (8) the sliding mode control of robot manipulator is calculated as;

$$U = \left[D^{-1}(f + C + G) + \dot{S}\right]D + K \cdot \text{SGN}(S)$$
(21)

Computed Torque Controller: The central idea of Computed torque controller (CTC) is feedback linearization so, originally this algorithm is called feedback linearization controller. It has assumed that the desired motion trajectory for the manipulator $q_d(t)$, as determined, by a path planner. Defines the tracking error as:

$$\boldsymbol{e}(t) = \boldsymbol{q}_d(t) - \boldsymbol{q}_a(t) \tag{22}$$

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I.J. Information Technology and Computer Science, 2014, 04, 66-80

Where e(t) is error of the plant, $q_d(t)$ is desired input variable, that in our system is desired displacement, $q_a(t)$ is actual displacement. If an alternative linear state-space equation in the form $\dot{x} = Ax + BU$ can be defined as

$$\dot{x} = \begin{bmatrix} 0 & I \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ I \end{bmatrix} U$$
(23)

With $U = -D^{-1}(q)$. $N(q, \dot{q}) + D^{-1}(q)$. τ and this is known as the Brunousky canonical form. By equation (22) and (23) the Brunousky canonical form can be written in terms of the state $x = [e^T \dot{e}^T]^T$ as [1]:

$$\frac{d}{dt} \begin{bmatrix} e \\ \dot{e} \end{bmatrix} = \begin{bmatrix} 0 & I \\ 0 & 0 \end{bmatrix} \cdot \begin{bmatrix} e \\ \dot{e} \end{bmatrix} + \begin{bmatrix} 0 \\ I \end{bmatrix} U$$
(24)

With

$$U = \ddot{q}_d + D^{-1}(q) \cdot \{N(q, \dot{q}) - \tau\}$$
(25)

Then compute the required arm torques using inverse of equation (25), is;

$$\tau = D(q)(\ddot{q_d} - U) + N(\dot{q}, q)$$
(26)

This is a nonlinear feedback control law that guarantees tracking of desired trajectory. Selecting proportional-plus-derivative (PD) feedback for U(t) results in the PD-computed torque controller [6];

$$\tau = D(q) \left(\ddot{q}_d + K_v \dot{e} + K_p e \right) + N(q, \dot{q})$$
(27)

and the resulting linear error dynamics are

$$\left(\ddot{q}_d + K_v \dot{e} + K_p e\right) = 0 \tag{28}$$

According to the linear system theory, convergence of the tracking error to zero is guaranteed [6]. Where K_p and K_v are the controller gains.

IV. Result and Discussion

PD-computed torque controller (CTC) and PD sliding mode controller (SMC) were tested to step and ramp responses. In this simulation, to control position of continuum robot manipulator the first, second, and third joints are moved from home to final position without and with external disturbance. The simulation was implemented in MATLAB/SIMULINK environment. Quality Factor Parameters of Control systems; trajectory performance, torque performance, disturbance rejection, steady state error and RMS error are compared in these controllers. These systems are tested by band limited white noise with a predefined 10%, 20% and 40% of relative to the input signal amplitude. This type of noise is used to external

disturbance in continuous and hybrid systems and applied to nonlinear dynamic of these controllers.

Tracking performances: based on formulation of computed torque controller; the performance of this controller is depended on the PD (K_p and K_v) coefficients. These two coefficients step and ramp CTC are $K_p = 30$ and $K_v = 4$ and computed by trial and error. Tables 1 and 2 shows the different coefficient gain in computed torque controller. (See table 1 and table 2)

Based on these tables, the different PD coefficient gain has the different errors therefore minimum error played important role to select the $K_p = 30$ and $K_v = 4$.

Based on sliding mode controller formulation with switching function; the performance is depended on the gain updating factor (*K*) and sliding surface slope coefficient (λ). These two coefficients are computed by trial and error based on Tables 3 and 4. (See table 3 and table 4)

Based on Table 3 and 4, in this research for step inputs $\lambda_1 = 3$ and $\lambda_2 = \lambda_3 = 6$; $K_1 = K_2 = K_3 = 30$ and for ramp inputs $\lambda_1 = \lambda_2 = 15$ and $\lambda_3 = 10$; $K_1 = K_2 = K_3 = 5$.

Based on sliding mode controller formulation with saturation function (boundary layer); the performance is depended on the gain updating factor (*K*), sliding surface slope coefficient (λ) and boundary layer saturation coefficient (\emptyset). These three coefficients are computed by trial and error based on Tables 5 and 6. (see table 5 and table 6)

In this research based on Tables 5 and 6, for step inputs: $\lambda_1 = 1, \lambda_2 = 6, \lambda_3 = 8; K_1 = K_2 = K_3 =$ 10; $\phi_1 = \phi_2 = \phi_3 = 0.1$ and for ramp inputs: $\lambda_1 =$ $\lambda_2 = 15, \lambda_3 = 10; K_1 = K_2 = K_3 = 5; \phi_1 = \phi_2 =$ $\phi_3 = 0.1$. Figures 3 and 4 are shown tracking performance for CTC, SMC with switching function and SMC with boundary layer without disturbance for two type trajectories.

Based on Figure 1; by comparing step response trajectory without disturbance in CTC, SMC with switching function and SMC with boundary layer, SMC with saturation function's overshoot about (0.94%) is lower than SMC with switching function's and CTC's (6.44%). SMC with switching function's and CTC's rise time (0.403) is lower than SMC with saturation function's (0.483). Based on simulation results the Steady State and RMS error = 1e-6 and RMS error=1.2e-6) are fairly lower than CTC's (Steady State error $\approx -3e - 5$ and RMS error=-1.34e - 5) and SMC with switching function's (Steady State error ≈ -0.001 and RMS error=0.00652). Based on Figure 1, it is found fairly fluctuations in trajectory responses.



Fig. 1: SMC with boundary layer, SMC with switching function and CTC for first, second and third joints: step trajectory

In Figure 2, all three controllers were tested for ramp trajectory. In this test, all three joints were required to move from home position (0) to a final position (+10) in 3.3 seconds. On reaching the final position the manipulator is stayed in another 26.7 seconds. This kind of trajectory is used to test the trajectory performance of these controllers. Based on chattering phenomenon challenge in pure sliding mode controller with switching function, it is found fairly fluctuations in trajectory responses.



Fig. 2: SMC with boundary layer, SMC with switching function and CTC for first, second and third joints: ramp trajectory

Based on Figure 1 and 2, step and ramp trajectory performances are used for comparisons of above controllers in certain systems. In this state CTC and SMC with saturation function has an acceptable trajectory performance but SMC with switching function has oscillation.

Disturbance rejection: Figures 3 to 8 show the power disturbance elimination in CTC, SMC with switching function and SMC with boundary layer without disturbance for two type trajectories. The disturbance rejection is used to test the robustness comparisons of these three controllers for ramp and step trajectories. A band limited white noise with predefined of 10%, 20% and 40% the power of input signal is

applied to the step and ramp trajectories. It found fairly fluctuations in trajectory responses.



Fig. 3: Desired input, SMC with boundary layer, SMC with switching function and CTC for first, second and third joints with 10% external disturbance: step trajectory

Based on Figure 3; by comparing step response trajectory with 10% disturbance of relative to the input signal amplitude in CTC, SMC with switching function and SMC with boundary layer, SMC with saturation function's overshoot about (1%) is lower than SMC with switching function's (8%)and CTC's (9.1%). SMC with switching function's and CTC's rise time (0.5) is lower than SMC with saturation function's (0.53). Besides the Steady State and RMS error in SMC with boundary layer (Steady State error =1.6e-6 and **RMS error=1.9e-6**) are fairly lower than CTC's (Steady State error **≅ 0.003** and RMS error=0.0048) and SMC with switching function's (Steady State error ≈ -0.1 and RMS error=0. 652). Based on Figure 3, SMC with switching function and CTC has oscillation in trajectory response with regard to 10% of the input signal disturbance but SMC with saturation function's trajectory is more robust.



Fig. 4: Desired input, SMC with boundary layer, SMC with switching function and CTC for first, second and third joints with 10% external disturbance: ramp trajectory

In Figure 4, all three controllers were tested for ramp trajectory. In this test, all three joints were required to move from home position (0) to a final position (+10) in

3.3 seconds in presence of 10% disturbance of input signal amplitude. On reaching the final position the manipulator is stayed in another 26.7 seconds. This kind of trajectory is used to test the trajectory performance of these controllers. Pure sliding mode controller with switching function and CTC has moderately fluctuations in trajectory responses. Based on Figures 3 and 4, step and ramp trajectory performances are used for comparisons of above controllers in presence of uncertainties with 10% of input signal amplitude. In this state SMC with saturation function has an acceptable trajectory performance but CTC and SMC with switching function have oscillation.



Fig. 5: Desired input, SMC with boundary layer, SMC with switching function and CTC for first, second and third joints with 20% external disturbance: step trajectory



Fig. 6: Desired input, SMC with boundary layer, SMC with switching function and CTC for first, second and third joints with 20% external disturbance: ramp trajectory

Based on Figure 5; by comparing step response trajectory with 20% disturbance of relative to the input signal amplitude in CTC, SMC with switching function and SMC with boundary layer, SMC with saturation function's overshoot about (2.1%) is lower than SMC with switching function's (9.4%) and CTC's (10.1%). SMC with switching function's and CTC's rise time (0.5) is lower than SMC with saturation function's (0.53). Besides the Steady State and RMS error in SMC

with boundary layer (Steady State error =1.8e-5 and RMS error=2e-5) are fairly lower than CTC's (Steady State error ≈ 0.002 and RMS error=0.022) and SMC with switching function's (Steady State error ≈ -0.15 and RMS error=0. 7). Based on Figure 5, however three controllers have oscillation in trajectory response with regard to 20% of the input amplitude signal disturbance but SMC with saturation function's trajectory is more robust.

In Figure 6, all three controllers were tested for ramp trajectory. In this test, all three joints were required to move from home position (0) to a final position (+10) in 3.3 seconds in presence of 20% disturbance of input signal amplitude. On reaching the final position the manipulator is stayed in another 26.7 seconds. This kind of trajectory is used to test the robust performance of these controllers. Pure sliding mode controller with switching function and CTC has fairly large fluctuations in trajectory responses but boundary layer SMC has relatively small oscillations. Based on Figures 5 and 6, step and ramp trajectory performances are used for comparisons of above controllers in presence of uncertainties with 20% of input signal amplitude. In this state boundary layer SMC has relatively small oscillations but it is more robust than CTC and SMC with switching function.



Fig. 7: Desired input, SMC with boundary layer, SMC with switching function and CTC for first, second and third joints with 40% external disturbance: step trajectory

Based on Figure 7; by comparing step response trajectory with 40% disturbance of relative to the input signal amplitude in CTC, boundary layer (saturation) SMC and switching mode SMC, SMC with saturation function's overshoot about (6%) is lower than SMC with switching function's (13%)and CTC's (14.8%). SMC with switching function's and CTC's rise time (0.5) is lower than SMC with saturation function's (0.53). Besides the Steady State and RMS error in SMC with boundary layer (Steady State error =10e-4 and RMS error=11e-4) are fairly lower than CTC's (Steady State error ≈ 0.005 and RMS error=0.075) and SMC with switching function's (Steady State error ≈ -0.98 and RMS error=0.98). Based on Figure 7, all three controllers have oscillation in

trajectory response with regard to 40% of the input signal amplitude disturbance.



Fig. 8: Desired input, SMC with boundary layer, SMC with switching function and CTC for first, second and third joints with 40% external disturbance: ramp trajectory

In Figure 8, all three controllers were tested for ramp trajectory. In this test, all three joints were required to move from home position (0) to a final position (+10) in 3.3 seconds in presence of 40% disturbance of input signal amplitude. On reaching the final position the manipulator is stayed in another 26.7 seconds. This kind of trajectory is used to test the robust performance of these controllers. All these three controllers have fairly large oscillation with regard to the external disturbance. Based on Figures 7 and 8, step and ramp trajectory performances are used for comparisons of above controllers in presence of uncertainties with 40% of input signal amplitude. In this state however boundary layer SMC has relatively moderate oscillations but it is more robust than CTC and SMC with switching function. From the disturbance rejection for boundary layer SMC, SMC with switching function and CTC in presence of disturbance, it was seen that however boundary layer SMC performance is better than SMC with switching function and CTC but it has a limitation against to highly external disturbance.

Steady state and RMS error: Figures 9 to 12 show the error performance without disturbance and in presence of disturbance in CTC, SMC with switching function and SMC with boundary layer without disturbance for two type trajectories. The error performance is used to test the disturbance effect comparisons of these three controllers for ramp and step trajectories. A band limited white noise with predefined of 40% the power of input signal is applied to the step and ramp trajectories. It found fairly fluctuations in error responses.

Figure 9 shows error performance for first three links of continuum robot manipulator in CTC, SMC with switching function and SMC with boundary layer without disturbance for step trajectory. Based on Figure 1, all three joint's inputs are step function with the same step time (step time= 1 second), the same initial value (initial value=0) and the same final value (final value=5). Based on Figure 3, all three controllers have about the same rise time (rise time=0.4 second) which it is caused to create a needle wave in the range of 5 (amplitude=5) and the time width of 0.6 second. In this system this time is transient time and this part of error is transient error. The SMC with boundary layer and CTC give considerable error performance when compared to SMC with switching function. Besides the Steady State and RMS error =1.2e-6) are fairly lower than CTC's (Steady State error $\approx -3e - 5$ and RMS error=-1.34e - 5) and SMC with switching function's (Steady State error ≈ -0.001 and RMS error=-0.00652).



Fig. 9: SMC with boundary layer, SMC with switching function and CTC for first, second and third joints steady state and RMS error without disturbance: step trajectory



Fig. 10: SMC with boundary layer, SMC with switching function and CTC First, second and third link steady state and RMS error with disturbance: step trajectory

Figure 10 shows steady state and RMS error performance for first three links of continuum robot manipulator in CTC, SMC with switching function and SMC with boundary layer with 40% disturbance for step trajectory. Based on Figure 1, all three joint's inputs are step function with the same step time (step time=1 second), the same initial value (initial value=0) and the same final value (final value=5). Based on Figure 1, all three controllers have about the same rise time (rise time=0.4 second) which it is caused to create

a needle wave in the range of 5 (amplitude=5) and the time width of 0.6 second. In this system this time is transient time and this part of error is transient error. The SMC with boundary layer gives considerable error performance when compared to CTC and SMC with switching function. Besides the Steady State and RMS error in SMC with boundary layer (Steady State error =10e-5 and RMS error=11e-4) are fairly lower than CTC's (Steady State error ≈ 0.005 and RMS error=0.075) and SMC with switching function's (Steady State error ≈ -0.98 and RMS error=0.98). Based on Figure 10, all three controllers have oscillation in error response with regard to 40% of the input signal amplitude disturbance. When applied 40% disturbances in SMC with boundary layer the RMS error increased approximately 9.17% (percent of increase steady the state no disturbance RMS error 1.2e - 6in CTC the RMS error increased approximately 56% the (percent of increase steady state $error = \frac{(40\% \ disturbance \ RMS \ error}{40\% \ disturbance \ RMS \ error}$ 0.075 = 56%). no disturbance RMS error 1.34e-5



Fig. 11: SMC with boundary layer, SMC with switching function and CTC First, second and third link steady state and RMS error without disturbance: ramp trajectory

Figure 11 shows steady state and RMS error performance for first three links of continuum robot manipulator in CTC, SMC with switching function and SMC with boundary layer without any disturbance for ramp trajectory. In Figure 11, all three controllers were

tested for ramp trajectory. In this test, all three joints were required to move from home position (0) to a final position (+10) in 3.3 seconds. On reaching the final position the manipulator is stayed in another 26.7 seconds. Based on Figure 11, CTC and SMC with boundary layer has acceptable performance in certain system without any disturbance but SMC with switching function has fairly oscillation in this state.



Fig. 12: SMC with boundary layer, SMC with switching function and CTC First, second and third link steady state and RMS error with disturbance: ramp trajectory

Figure 12 shows steady state and RMS error performance for first three links of continuum robot manipulator in CTC, SMC with switching function and SMC with boundary layer with 40% disturbance for ramp trajectory. In Figure 14, all three controllers were tested for ramp trajectory. In this test, all three joints were required to move from home position (0) to a final position (+10) in 3.3 seconds. On reaching the final position the manipulator is stayed in another 26.7 seconds. Based on Figure 12, in presence of 40% disturbance however all of three controllers have oscillation but SMC with boundary layer has better performance compared to CTC and SMC with switching function. This is mainly because SMC with saturation function is more robust than CTC and SMC with switching function.

Chattering phenomenon: Chattering is one of the most important challenges in sliding mode controller therefore reducing the chattering is a major objective in

this research. Chattering phenomenon is caused to the hitting in driver and mechanical parts therefore reduce the chattering is very important in this research. To reduce the chattering this research is focused on *saturation* function instead of *switching* function. Figure 13 shows the power of boundary layer (saturation) method to reduce the chattering in sliding mode controller.



Fig. 13: SMC with boundary layer Vs SMC with switching function: chattering phenomenon

V. Conclusion

Refer to this research, a comparative study between position sliding mode controller and computed torque controller is proposed for OCTARM continuum robot manipulator. The first problem of the pure sliding mode controller with switching function was chattering phenomenon in certain and partly uncertain systems. The chattering phenomenon problem can be reduced in certain system by using linear saturation boundary layer function in sliding mode control law. The simulation results exhibit that the sliding mode controller with saturation function works well in certain and partly uncertain system. Computed torque controller is another type of nonlinear controller. This controller is worked based on feedback control law and compute the required torque arm based on equivalent part. This methodology is based on applied nonlinear dynamic formulation in linear behaviour part to estimate nonlinear term parameters. The results demonstrate that the sliding mode controller with saturation function is a model-based controllers which works well in certain and partly uncertain system but computed torque

controller is an essential controller for certain nonlinear plant.

Acknowledgment

The authors would like to thank the anonymous reviewers for their careful reading of this paper and for their helpful comments. This work was supported by the Institute of Advanced Science and Technology (IRANSSP) Research and Development Corporation Program of Iran under grant no. 2013-**Persian Gulf**-2B.

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Appendixes

	k_{P_1}	k_{V_1}	k_{p_2}	k_{V_2}	k_{p_3}	k_{V_3}	RMs error	SS error ₁	SS error 2	SS error ₃
1	8	4	8	4	8	4	2.276e-5	-3.81e-5	-3.81e-5	-3.81e-5
2	30	4	30	4	30	4	1.34e-5	-3.6e-5	-2.54e-5	-1.6e-5
3	1	4	1	4	1	4	0.0039	0.0065	0.0065	0.0065
4	8	40	8	40	8	40	0.502	5.043	5.043	5.043
5	8	0.5	8	0.5	8	0.5	0.0026	0.0043	0.0043	0.0043

 Table 1: Tuning parameters of a step PD-CTC

Table 2:	Tuning	parameters	of a	ramp	PD-CTC
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	k_{P_1}	k_{V_1}	k_{p_2}	k_{V_2}	k_{p_3}	k_{V_3}	RMs error	SS error ₁	SS error 2	SS error ₃
1	8	4	8	4	8	4	32e-5	-28.1e-5	-26e-5	-21e-5
2	30	4	30	4	30	4	-6e-5	-8e-5	-8.6e-5	-8.9e-5
3	1	4	1	4	1	4	0.000305	0.00024	0.00024	0.00024
4	8	40	8	40	8	40	0.6	4.93	4.93	4.93
5	8	0.5	8	0.5	8	0.5	0.5	2.9	2.9	2.9

_	λ_1	k_1	λ_2	k_2	λ_3	<i>k</i> ₃	SS error ₁	SS error 2	SS error ₃	RMS error
data1	3	30	6	30	6	30	0.1e-3	0.1e-3	-5.3e-15	0.1e-4
data2	30	30	60	30	60	30	-5.17	14.27	-1.142	0.05
data3	3	300	6	300	6	300	2.28	0.97	0.076	0.08

Table 3: Tuning parameters of Step SMC with switching function

Table 4: Tuning parameters of a Ramp SMC with ramp switching function

	λ_1	k_1	λ_2	<i>k</i> ₂	λ_3	<i>k</i> ₃	SS error ₁	SS error 2	SS error ₃	RMS error
data1	15	5	15	5	10	5	4.6e-12	-3.97e-12	-3.87e-12	0.0002441
data2	150	5	150	5	100	5	1005	1108	436.5	0.8
data3	15	50	15	50	10	50	-0.1877	-0.1	-0.03	0.0006579

Table 5: Tuning parameters of a step SMC with boundary layer

	k_1	λ_1	ϕ_1	k_2	λ_2	ϕ_2	k_3	λ_3	ϕ_3	SS error ₁	SS error 2	SS error ₃	RMS error
data1	10	1	0.1	10	6	0.1	10	8	0.1	1e-6	1e-6	1e-6	1.2e-6
data2	100	1	0.1	100	6	0.1	100	8	0.1	0.2	0.05	-0.02	-0.037
data3	10	10	0.1	10	60	0.1	10	80	0.1	0.22	-0.21	-0.19	0.09

Table 6: Tuning parameters of a ramp SMC with boundary layer

	k_1	λ_1	ϕ_1	k_2	λ_2	ϕ_2	<i>k</i> ₃	λ_3	\$ _3	SS error ₁	SS error 2	SS error ₃	RMS error
data1	5	15	0.1	5	15	0.1	5	10	0.1	-6e-12	-8.5e-11	-1.7e-11	8.3e-5
data2	50	15	0.1	50	15	0.1	50	10	0.1	0.09	0.06	0.02	0.00162
data3	5	150	0.1	5	150	0.1	5	100	0.1	377.7	377	272	0.732

Authors' Profiles



Alireza Salehi is currently working as a primary researcher in the laboratory of Control and Robotic, Institute of Advance Science and Technology, IRAN SSP research and development Center. His current research interests are in the area of nonlinear control, artificial control system and robotics.

Farzin Piltan was born on 1975, Shiraz, Iran. In 2004 he is jointed the research and development company, SSP Co, Shiraz, Iran. In addition to 7 textbooks, Farzin Piltan is the main author of more than 95 scientific papers in refereed journals. He is

supervised and managed nonlinear control and artificial intelligent Lab, Institute of Advance Science and Technology for over 3 years. He also has highly skilled in embedde system controller and device programming including FPGA and CPLD. He is editorial review board member for 'international journal of control and automation (IJCA), Australia, ISSN: 2005-4297; 'International Journal of Intelligent System and Applications (IJISA)', Hong Kong, ISSN:2074-9058; 'IAES international journal of robotics and automation, Malaysia, ISSN:2089-4856; 'International Journal of Reconfigurable and Embedded Systems', Malaysia, ISSN:2089-4864. His current research interests are nonlinear control, artificial control system and applied to FPGA, robotics and artificial nonlinear control and IC engine modeling and control.



Mahdi Mirshekaran is currently working as a primary researcher in the laboratory of Control and Robotic, Institute of Advance Science and Technology, IRAN SSP research and development Center. His current research interests are in the area of nonlinear control, artificial control

system and robotics.



Meysam Kazeminasab is currently working as a primary researcher in the laboratory of Control and Robotic, Institute of Advance Science and Technology, IRAN SSP research and development Center. His current research interests are in the area of nonlinear control, artificial control

system and robotics.



Zahra Esmaeili is currently working as a primary researcher in the laboratory of Control and Robotic, Institute of Advance Science and Technology, IRAN SSP research and development Center. Her current research interests are in the area of

nonlinear control, artificial control system and robotics.

How to cite this paper: Alireza Salehi, Farzin Piltan, Mahdi Mirshekaran, Meysam Kazeminasab, Zahra Esmaeili, "Comparative Study between Two Important Nonlinear Methodologies for Continuum Robot Manipulator Control", International Journal of Information Technology and Computer Science(IJITCS), vol.6, no.4, pp.66-80, 2014. DOI: 10.5815/ijitcs.2014.04.08