

Design and Implementation of Fuzzy Controller on FPGA

Mani Shankar Anand

Department of Electrical Engineering, Indian Institute of Technology, Roorkee, India manishankar.anand@yahoo.com

Barjeev Tyagi

Department of Electrical Engineering, Indian Institute of Technology, Roorkee, India btyagfee@iitr.ernet.in

Abstract—Fuzzy Logic Controller (FLC) systems have emerged as one of the most promising areas for Industrial Applications. The highly growth of fuzzy logic applications led to the need of finding efficient way to hardware implementation. Field Programmable Gate Array (FPGA) is the most important tool for hardware implementation due to low consumption of energy, high speed of operation and large capacity of data storage. In this paper, instead of an introduction to fuzzy logic control methodology, we have demonstrated the implementation of a FLC through the use of the high speed integrated circuits Hardware Very Description Language (VHDL) code. FLC is designed for an armature control DC motor speed control. VHDL has been used to develop FLC on FPGA. A Sugeno type FLC structure has been used to obtain the controller output. The controller algorithm developed synthesized, simulated and implemented on FPGA Spartan 3E xc3s500e-4fg320 board.

Index Terms—FLC, VHDL, Hardware Implementation, FPGA

I. Introduction

The past few years have witnessed a rapid growth in the number and variety of application of fuzzy logic. The application ranges from consumer products such as cameras, washing machines, cars and in industry for medical instrumentation, underground trains and robots. Unlike the conventional controller FLC design is not based on the mathematical model of the plant or system. A FLC is an automatic controller that controls an object in accordance with desire behaviour. For a complex system whose mathematical model is very difficult to define or the transfer function of a plant is undefined, fuzzy logic controllers are very useful in that case [1,3]. The control action of FLC is defined in terms of simple human friendly "if - then rules". These set of rules are describe the system behaviour. These set of rules are called the knowledge base of fuzzy controller. We can easily change the rules accordance with our desire output. So the development time for a new controller can be significantly reduced as compared to conventional one [7]. The motivation behind the

implementation of a FLC in VHDL was driven by the need for an inexpensive hardware implementation of a generic fuzzy controller for use in industrial and commercial applications [13]. We have taken a simple FLC for an armature control DC motor speed control. Error and change in error in speed has been used as two inputs to FLC. For both the inputs 5 triangular membership function has been selected and coded in VHDL. An algorithm has been developed in VHDL to fuzzifie the crisp digital values of speed error and rate of change of error. Sugeno type FLC structure has been used to obtain the controlled output. The controller algorithm developed synthesized, simulated and implemented on FPGA Spartan 3E xc3s500e-4fg320 board. The FLC has been design using system generator approach. The results of the FLC implemented on FPGA have been compared with the results obtained using PID and FLC on MATLAB Simulink.

II. System Description

DC servomotors used in many applications, such as steel rolling mills, electric trains, and robotic manipulators, require speed controllers to perform tasks [4]. There are numerous control algorithms; however, the PID control is the most popular because of its simplicity. Major problems in applying a conventional control algorithm in a speed controller are the effects of nonlinearity in a DC motor. The nonlinear characteristics of a DC motor such as saturation and friction could degrade the performance of conventional controller. Every control project starts with plant modelling, so as much information as possible is given about the process itself. The mechanical-electrical model of the servomotor is presented in Fig.1.



According to the electrical-mechanical diagram presented in Figure 1, the linear model can be derived by the following dynamics governing equations:

$$\tau = K_t i \tag{1}$$

$$V_{emf} = K_b \dot{\theta}$$
(2)

$$J\dot{\theta} = -f\dot{\theta} + K_t i \tag{3}$$

$$E - V_{emf} = -L\frac{di}{dt} + Ri \tag{4}$$

Equations (1) to (4) constitute the two main differential equations. Their transformation leads to state space form:

$$\begin{pmatrix} \ddot{\theta} \\ \dot{i} \end{pmatrix} = \begin{pmatrix} -f / J & K_i / J \\ -K_b / L & -R / L \end{pmatrix} \begin{pmatrix} \dot{\theta} \\ \dot{i} \end{pmatrix} + \begin{pmatrix} 0 \\ 1 / L \end{pmatrix} E$$

$$y = \begin{pmatrix} 1 & 0 \end{pmatrix} \begin{pmatrix} \dot{\theta} \\ \dot{i} \end{pmatrix}.$$
(5)

From Eq. (5), it is clear that the DC motor is a SISO plant; however, the output choice of the model changes on the control scheme, we adopt, i.e., either velocity or position control.

The closed-loop transfer function of the DC motor can be calculated using Eq. (5) as

$$G_{v}(s) = \frac{w(s)}{e(s)} = \frac{K_{t}}{JLs^{2} + (JR + fL)s + fR + K_{b}K_{t}}$$
(6)

where E is the armature voltage of a dc motor, state variable $x = [w i]^T$, where i and w are the armature current and shaft rotational speed respectively, All motor parameter are obtained by standard system identification as shown in Table 1[12]

Parameter	Description	Value
R	Armature resistant	4.67 Ω
L	Armature inductance	170e-3 H
J	Moment of inertia	42.6e-6 Kg-m ²
f	Viscous-friction coefficient	47.3e-6 N- m/rad/sec
Kt	Torque constant	14.7e-3 N-m/A
K _b	Back-EMF constant	14.7e-3 V-sec/rad

Table I DC Motor parameters

Putting the value of all parameter we can get the transfer function

$$G_{\nu}(s) = \frac{14700}{7.242s^2 + 206.983s + 436.981} \tag{7}$$

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III. Implementation of FLC in VHDL

The widely used method for Fuzzy Logic is Mamdani style because of its simplicity and its compatibility with human made decision. Although Mamdani Style is widely used, it is not suitable for this implementation. This is because, the Mamdani method requires finding the centroid of a two-dimensional shape by integrating across a continuously varying function. This method is not computationally efficient. It is also not suitable because its output use triangle or trapezoidal membership function to describe the output and its defuzzification calculation is complicated to be done in VHDL despite the results are not necessary effective or better[7]. Sugeno style is much more suitable to be used in this situation. This is because, the output of this method, uses a single spike, singletons to describe the outputs where there is a unity at single particular point and zero elsewhere. This causes all the output of each fuzzy rule to be constant [8].

1) Fuzzification

The first component in the FLC is the fuzzifier that converts crisp inputs into a set of membership values in the interval [0,1] in the corresponding fuzzy sets. In this paper, triangular membership functions (triangular membership function being a special case of the trapezoidal function) are used for two inputs Error and change in error in speed [8]. Each of inputs is represented by 5 membership functions, which are NB (Negative Big), NM (Negative Medium), ZE (Zero), PM (Positive Medium), PB (Positive Big) as shown in Fig.2 and Fig.3.



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In fuzzification, the degree of membership function is determined. This is done by locating the location of the input in the membership function and determines the degree of the membership function. The degree of the membership function is from 0 to 1 where in this paper, it is represented by X00 and XFF ("X" sign indicates hexadecimal number representation).

In VHDL, Each trapezoidal membership function is defined by two points and two slopes values. The entire membership function can be divided into three segments: I, II and III as shown in Fig4. The Y axis shows the degree of membership function (μ) as a value between 0 and 1. The X axis shows the universe of discourse and is divided into three segments. The degree of membership depends on the location of the input value with reference to these segments. Fig4 shows how trapezoidal input membership functions are formed in the fuzzification process [8].



Fig.4. Trapezoidal Type Membership Function

The calculation of the degree of membership (μ) can be categorized into three different segments: (a) in segments I: $\mu = 0$, (b) in segments II: slope is upward from left to right, therefore: $\mu = (Input value - point 1)$ * slope1, where μ is limited to a maximum value of 1, (c) in segments III: slope is downward from left to right, therefore: $\mu = 1$ - (Input value - point 2) * slope2, where μ is limited to a minimum value of 0. The value of slope1 and slope2 can be calculated as follows:

Slope
$$1 = 1 / (b - a) = XFF / (b - a) = 255 / (b - a)$$
 (8)

Slope
$$2 = 1 / (d - c) = XFF / (d - c) = 255 / (d - c)$$
 (9)

Declaration of membership functions in VHDL as follows:

type input is (term, none); type mfs is record linguistic: input; point1: std_logic_vector (7 downto 0); slope1: std_logic_vector (7 downto 0); point2: std_logic_vector (7 downto 0); slope2: std_logic_vector (7 downto 0); end record; $\begin{array}{l} type \ mfs_functions \ is \ array \ (natural \ range <>) \ of \ mfs; \\ constant \ linguistic_name: \ mfs_functions:= \\ ((linguistic => term, \ point1 => x"04", \ slope1 => x"7F", \\ point2 => x"09", \ slope2 => x"55"), \\ (linguistic => \ none, \ point1 => x"FF" \ , \ slope1 => \\ x"FF", \ point2 => x"FF", \ slope2 => x"FF")); \end{array}$

After declaration of membership functions we have to calculate the degree of membership function. Since a specific input value only intersect at most two membership functions to create the degree of membership function for the corresponding input, so most of them will be zero.

The following algorithm illustrates the procedure of this process:

 $\begin{aligned} &Fuzzification ();\\ &Set n = number of membership function\\ &\mu = array of degree of membership function\\ &m = array of membership function\\ &start loop for i = 1 to n\\ & if input value < m[i].point1 then \mu[i]=0;\\ &else if input value < m[i].point2 then \mu[i]=(input value-m[i].point1) \times m[i].slope1;\\ &else \ \mu[i]= 255- (input value - m[i].point2)\\ &\times m[i].slope2;\\ &end if;\\ &end loop;\end{aligned}$

2) Rule Inference

The degree of membership is determined in the fuzzification stage. The next step is to create rules to decide what action should be taken in response to the given set of degree of membership function. The "AND" and "OR" fuzzy operators are best used for rules with multiple antecedents. The fuzzy operator, "OR" is used to evaluate the disjunction of the rules antecedents and "AND" is used to evaluate the conjunction of the rules antecedents. "AND" fuzzy operator is since it is required to evaluate the conjunction of the rules antecedents. Since "AND" is the minimum operation between multiple antecedents, the minimum function is used. The "OR" fuzzy operator also can be used when more than one rules involved with the same output. The rule base of systemis defined in Table 2[12].

Table 2 Fuzzy final rules

ECE	NB	NM	ZE	PM	PB
PB	NM	NS	NB	PB	PB
PM	NM	NM	NB	PB	PB
ZE	NB	NB	ZE	PB	PB
NM	NB	NB	NB	PM	PM
NB	NB	NB	NB	PS	PM

Below is the example of implementation of rules and the minimum and maximum;

For "OR" =>	C = maximu m	(A, B)
For "AND" =>	C = minimum	(A, B)

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For rules with same output =>

C = maximum (minimum (A1, B1), minimum (A2, B2))

The minimum and maximum function used to obtain the result from the each rule evaluation between multiple antecedents is shown below.

Minimum Function

function minimum (a, b : std_logic_vector) retum
std_logic_vectoris
variable min : std_logic_vector (7 downto 0) := (others
=> '0');
begin
 if (a < b) then min := a;
 elsif min := b;
end if;
return min;
end minimum;</pre>

Maximum function

function maximum (a, b : std_logic_vector) retum
std_logic_vector isvariable max : std_logic_vector (7
downto 0) := (others => '0');
begin

```
if (a > b) then max := a;
elsif max := b;
end if;
return max;
end maximum;
```

3) Rule Evaluation

Rule 1: position(1)<=ma ximu m(minimu m (u1(0), u2(1)), minimu m (u1(1), u2(0)));

A total number of rules that should be produced to describe the complete fuzzy control strategy can be calculated by multiplying the input membership function with the output membership function. Although there are number of possible rules, most of them can be discarded as long as the design is able to determine how the fuzzy control system should be operated.

4) Defuzzification

After the output for the each rule has been identified, the next step is to combine all the output into a single value that can used to control the motors. This process is done through defuzzification. The defuzzification technique used in Sugeno method is weighted average. This is done by multiplying fuzzy output obtained from the rules evaluation with its corresponding singleton value, then sum of this value is divided by the sum of all fuzzy output obtained from the rules evaluation. The result from this calculation is the final single output which can be used to control the motor movements. Since there is no division symbol supported by Xilinx ISE Compiler, a divider circuit has to be designed to perform defuzzification.

The following pseudo-code illustrates the procedure of this fuzzification process [8]:

Defuzzification();

Set *n* = number of output membership function

s = array of singleton of output membership function

p = array of result of all rule evaluation sum = 0;For i = 1 to n do begin product = $(s(i) \times p(i)) + product;$

sum = p(i) + sum; end for loop; output = product / sum;

IV. Simulation and Results

Simulation in MATLAB

Simulations have been done for DC motor using FLC in MATLAB/SIMULINK. The simulink diagram of FLC is shown in Fig.5. The comparisons of FLC has been done with PID controller expressed by

$$c(s) = K_p + \frac{K_i}{s} + K_d s$$

Where Kp, Kd, Ki are 1, 0.6, 0.5 respectively. The PID controller is tuned using Zeigler-Nichols method. The step response using FLC and PID are shown in Fig.6. The FLC brings fastest settling and less overshoot in compared to PID. Hence the performance of FLC is better than PID



Fig.5. Simulink Diagram of FLC and PID



Fig.6. Step response of system with feedback FLC

Simulation in Xilinx ISE

Simulations have been done in Xilinx ISE.Fig.7 and Fig.8 shows that Test Bench Waveform and RTL view of FLC using VHDL respectively. In this figure the outputs have calculated according to input i.e. Error and Change in Error

Current Simulation				950.0) ne
Time: 1000 ns		400 ns 500 ns 60	0 ns 700 ns 8	00 ns 900 ns10	00
👌 clock	1				_
🗉 🛃 in1 [7:0]	8'hE0	8'hE0 X	8'hA0	(8"hED	
🗉 💦 in2[7:0]	8'h60	8°h60 X	B'hEO	8"h60	
🗄 😹 product[15:0]	1	(16'h142F	16'h9CDE	16'h142F	
🖽 💦 sum[15:0]	1	(16'h9DFA	16'hB73A	X 16'h9DFA	
🛃 period	2	20000000			
👌 duty_cycle	0.5	0.5			
👌 oπset	1		100000000		

Fig.7. Test Bench Waveform using Xilinx ISE



Fig.8. RTL view of FLC using VHDL

Fig.9 shows the Simulink Diagram of FLC and PID using Xilinx system generator. We have taken a same 2^{nd} order system for Dc motor to show the step response performance of FLC. Black box represents the FLC using VHDL code. The two input of FLC i.e. Error and Change in Error are going through the Gateway In to the system generator Black Box, the two outputs are coming through the Black Box. The control output of FLC is finally obtained by dividing one output by another.



Fig.9. Simulink Diagram of FLC and PID using Xilinx system generator

Fig. 10 shows the Comparisons among the FLC, FLC using VHDL and PID controller. FLC using VHDL the lowest settling time, steady state error and less

overshoot in compared to PID and FLC. Hence the performance of FLC using VHDL is better than PID and FLC



Fig.10. Comparisons among PID, FLC and FLC-VHDL

Fig.11 shows that Simulink diagram after hardware co-simulation. Whole Xilinx block set is represented by JTAG co-sim block.



Fig.11. Simulink diagram after hardware co-simulation

Fig.12 verifies that the response of FLC after hardware co-simulation is almost same as compare to FLC before hardware co-simulation.



Fig.12. Comparisons between PID controller and FLC-VHDL after hardware co-simulation

V. Conclusion

For designing FLC, a high-level modelling approach in VHDL have to be used. The advantages of this are reducing the design time, evaluation of the design functionality in a short time and quickly exploring of different design choices. Once the basic design of the fuzzy logic control system has been defined, the implementation of the fuzzy logic controller is very straight forward by coding each component of the fuzzy inference system in VHDL according to the design specifications. By simply changing some parameters in the codes and design constraint on the specific synthesis tool, one can experiment with different design circuitry to get the best result in order to satisfy the system requirement. From the results, we saw that the response of FLC using VHDL is better than the response of PID and FLC using MATLAB. The peak overshoot and settling time both are better than others.

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Mani Shankar Anand received the M.Tech Electrical Engineering (System & Control) from IIT-Roorkee in the year 2012. Prior to these he completed his B.Tech Electronics & Communication Engineering from

F.I.E.M. Kolkata in 2010. His research interests include Nonlinear control.



Barjeev Tyagi received the PhD in Electrical Engineering, IIT-Kanpur, 2005 and M.Tech Electrical Engineering (Control System) from IIT-Kharagpur in the year 2000. Prior to these he completed his B. E. Electrical Engineering from IIT-

Roorkee (Formally Univ. of Roorkee) in 1987. Presently he is serving as Assistant Professor in EED, IIT-Roorkee since 2007. His research interests include power system deregulation, power system optimization, distributed generation and control.

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