Nonlinear Blind Source Separation Using Kernel Multi-set Canonical Correlation Analysis

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Abstract—To solve the problem of nonlinear blind source separation (BSS), a novel algorithm based on kernel multi-set canonical correlation analysis (MCCA) is presented. Combining complementary research fields of kernel feature spaces and BSS using MCCA, the proposed approach yields a highly efficient and elegant algorithm for nonlinear BSS with invertible nonlinearity. The algorithm works as follows: First, the input data is mapped to a high-dimensional feature space and perform dimension reduction to extract the effective reduced feature space, translate the nonlinear problem in the input space to a linear problem in reduced feature space. In the second step, the MCCA algorithm was used to obtain the original signals.

Index Terms—nonlinear blind source separation, kernel feature spaces; multi-set canonical correlation analysis; reduced feature space; joint diagonalization

I. INTRODUCTION

In recent years, blind source separation (BSS) raises great interest. In fact, BSS plays an important role in many diverse application areas, such as radio communications, radar, sonar, seismology, image processing, speech processing (cocktail party problem) and biomedical signal analysis where multiple sensors are involved. In the last years, linear BSS has become relatively well established signal processing and data analysis techniques [1]. On the other hand, nonlinear BSS is technique that is still largely under development, and has the potential to become rather powerful tools [2].

Kernel-based methods have also been considered for solving the nonlinear BSS problem [3], [4]. The data are first implicitly mapped to high-dimensional feature space, and the effective reduced feature space in feature space is extracted, translate the nonlinear problem in the input space to a linear problem in reduced feature space.

Canonical correlation analysis (CCA) is a classical tool in multivariate statistical analysis to find maximally correlated projections between two data sets, since it was proposed by H. Hotelling [5]. There are detail descriptions of CCA in [6], which has been widely used in many modern information processing fields, such as for test of independence [7], blind equalization of MIMO channels [8] and BSS [9],[10]. With CCA, the objective is to find a transformation matrix which is applied to mixtures and maximizes the autocorrelation of each of the recovered signals (the outputs of the transformation matrix). By maximizing this autocorrelation, the original uncorrelated source signals will be recovered. This approach rests on the idea that the sum of any uncorrelated signals has an autocorrelation whose value is less or equal to the maximum value of individual signals [11]. The algorithm based on CCA which computation burden is little and the flexibility is strong, can satisfy the demand of engineering application. Based on the idea of maximize generalized relativity measurement, J. Kettingring present-ed several ways to generalize CCA to more than two sets of variables, namely MCCA (Multi-set CCA) [12]. By assuming that sources are not correlated with the others and every source has a different temporal structure, which is a mild condition that can be easily, satisfied in practical applications, a MCCA linear BSS algorithm was proposed in [13]. This paper proposed a novel nonlinear BSS algorithm based on kernel feature spaces and MCCA. The algorithm can adapt to nonlinear BSS with invertible nonlinearity. The details of the new method will be described in the following sections. The paper is organized as follows: Section II presents the detail analysis of linear BSS algorithm based on MCCA. In the Section III, the new nonlinear BSS algorithm based on kernel MCCA will be analyzed. Then some simulations of the algorithm proposed in this paper are conducted in the Section IV. Section V is the conclusion.

II. LINEAR BSS ALGORITHM BASED ON MCCA

A. Linear BSS problem formulation

Consider the following instantaneous linear mixture model:

\[ x(t) = Hs(t) + n(t) \]

where \( x(t) = (x_1(t), x_2(t), \ldots, x_m(t))^T \) is the \( m \)-dimensional vector of mixed signals observed by \( m \) sensors, \( n(t) \) is the additive noise vector, \( H \) is the unknown \( m \times m \) mixing matrix, \( s(t) = (s_1(t), s_2(t), \ldots, s_m(t))^T \) is the \( n \)-dimensional vector of source signals (which is also unknown and \( m > n \)), and the superscript \( T \) denotes the transpose operator. BSS methods aim at estimating the source signals \( s(t) \).

Suppose the source signals are not correlated with the others and every source has a different temporal structure, and the additive noise vector \( n(t) \) is statistically independent of \( s(t) \).
The task of BSS is to estimate the mixing matrix \( H \) (or is pseudo-inverse, \( W = H^T \)) that is referred to as the demixing matrix), given only a finite number of observation data \( x(t) \) \( t = 1, T \). And obtain the estimated source signals \( y(t) = (y_1(t), y_2(t), \ldots, y_L(t))^T \) \( t = 1, L \).

\[
y(t) = W x(t). \tag{2}
\]

We just consider the case of \( m = n \), for less sources than mixtures (\( m > n \)) the BSS problem is said to be over-determined, and it is easily reduced to a square BSS problem by selecting \( m \) mixtures or applying some more sophisticated preprocessing like PCA.

### B. MCCA

CCA is a multivariate statistical technique similar in spirit to principal component analysis (PCA). While PCA works with a single random vector and maximizes the variance of projections of the data, CCA works with a pair of random vectors (or in general with a set of \( m \) random vectors) and maximizes correlation between sets of projections.

Given two random vectors, \( x_1 \) and \( x_2 \) of dimension \( p_1 \) and \( p_2 \). The first canonical correlation can be defined as the maximum possible correlation between the two projections \( u_i = a_i^{(1)} x_i \) and \( v_i = a_i^{(2)} x_i \) of \( x_1 \) and \( x_2 \) [6]:

\[
\rho_1(x_1, x_2) = \max_{a_i^{(1)}, a_i^{(2)}} \text{corr}(a_1^{(1)} x_1, a_1^{(2)} x_2) = \max_{a_i^{(1)}, a_i^{(2)}} \frac{a_i^{(1)T} C_1 a_i^{(1)} a_i^{(2)T} C_2 a_i^{(2)}}{(a_i^{(1)T} C_1 a_i^{(1)})^{1/2} (a_i^{(2)T} C_2 a_i^{(2)})^{1/2}} \tag{3}
\]

where \( C_1 = x_1 x_1^T, C_2 = x_2 x_2^T, C_{11} = x_1 x_1^T, C_{22} = x_2 x_2^T \).

After finding the first pair of optimal vectors \( a_i^{(1)} \) and \( a_i^{(2)} \), we can proceed to find the second pair \( a_i^{(1)} \) and \( a_i^{(2)} \) which maximizes the correlation and at the same time ensures that the new pair of combinations \( \{u_i, v_i\} \) is uncorrelated with the first set \( \{u_i, v_i\} \). This process is repeated until we find all \( p_1 \) and \( p_2 \) pairs of optimal vectors \( a_i^{(1)} \) and \( a_i^{(2)} \), \( i = 1, 2, \ldots, p \).

Since the choice of rescaling is therefore arbitrary, Normalizing the vectors \( a_i^{(1)} \) and \( a_i^{(2)} \) by letting \( a_i^{(1)T} C_1 a_i^{(1)} = 1 \) and \( a_i^{(2)T} C_2 a_i^{(2)} = 1 \), we see that CCA reduces to the following Lagrangian [6]:

\[
L(\lambda, a_i^{(1)}, a_i^{(2)}) = a_i^{(1)T} C_1 a_i^{(1)} - \frac{\lambda^{(1)}}{2} (a_i^{(2)T} C_1 a_i^{(2)} - 1) \tag{4}
\]

Taking derivatives in respect to \( a_i^{(1)} \) and \( a_i^{(2)} \), the CCA problem can be obtained by solving the following generalized Eigen values problem

\[
C_i C_j a_i^{(1)} = \lambda^{(1)} C_i a_i^{(1)}, \quad C_i C_j a_i^{(2)} = \lambda^{(2)} C_2 a_i^{(2)}, \tag{5}
\]

There’s a generalized Eigen problem of the form \( Ax = \lambda Bx \). We can therefore find the coordinate system that optimizes the correlation between corresponding coordinates by first solving for the generalized eigenvectors of (5) to obtain \( a_i^{(1)} \) and \( a_i^{(2)} \).

According to the ways in [6] we can generalize CCA problem to the minimization of the total distance.

For CCA, two matrices \( A^{(1)} \) and \( A^{(2)} \) contain the vectors \( \{a_1^{(1)}, \ldots, a_p^{(1)}\} \) and \( \{a_1^{(2)}, \ldots, a_p^{(2)}\} \). The CCA problem can also be express as

\[
\max_{\lambda^{(1)}, \lambda^{(2)}} \text{tr}(A^{(1)T} C_1 A^{(1)} A^{(2)T} C_2 A^{(2)})
\]

s.t. \( A^{(1)T} C_1 A^{(1)} = I, \quad A^{(2)T} C_2 A^{(2)} = I \)

(6)

\[
a_i^{(1)T} C_1 a_i^{(1)} = 0, \quad a_i^{(2)T} C_2 a_i^{(2)} = 0, \quad i = 1, \ldots, p \tag{6}
\]

where \( I \) is an identity matrix with size \( q \times q \).

The canonical correlation problem can be transformed into a distance minimization problem where the distance between two matrices is measured by the Frobenius norm:

\[
\min_{\lambda^{(1)}, \lambda^{(2)}} \| x_i A^{(1)} - x_i A^{(2)} \|_F
\]

s.t. \( A^{(1)T} C_1 A^{(1)} = I, \quad A^{(2)T} C_2 A^{(2)} = I \)

(7)

\[
a_i^{(1)T} C_1 a_i^{(1)} = 0, \quad a_i^{(2)T} C_2 a_i^{(2)} = 0, \quad i = 1, \ldots, p \tag{7}
\]

We can give a similar definition for MCCA. Given \( m \) multivariate random variables in matrix form \( \{x_1, \ldots, x_k\} \).

We are looking for the linear combinations of the columns of these matrices in the matrix form \( \{A^{(1)}, \ldots, A^{(k)}\} \) such that they give the optimum solution of the problem:

\[
\min_{\lambda^{(1)}, \lambda^{(2)}} \sum_k \sum_i \| x_i A^{(k)} - x_i A^{(i)} \|_F
\]

s.t. \( A^{(1)T} C_1 A^{(1)} = I, \quad A^{(2)T} C_2 A^{(2)} = I \)

(8)

\[
a_i^{(1)T} C_1 a_i^{(1)} = 0, \quad a_i^{(2)T} C_2 a_i^{(2)} = 0, \quad k, l = 1, \ldots, K, i \neq k, i, j = 1, \ldots, p, j \neq i \]

which is the sum of the squared Euclidean distances between all of the pairs of the column vectors of the matrices \( x_i A^{(k)}, k = 1, \ldots, K \).

### C. BSS Algorithm Analysis

We start by specifying the signal. It is assumed that the sources are spatially uncorrelated, the correlation matrix of the sources \( R_s(\tau) = E\{s(t) s^T(t+\tau)\} \) is a diagonal matrix, where \( E\{\} \) denotes the statistical expectation operator. For nonzero correlation lags, we have

\[
C_{ij} C_{kl} a_i^{(1)} = \lambda^{(1)} C_{ij} a_i^{(1)}, \quad C_{ij} C_{kl} a_i^{(2)} = \lambda^{(2)} C_{22} a_i^{(2)}.
\]
\( R_\alpha[\tau] = E\{ s(t)s^T(t-\tau) \} = \text{diag}\{ \lambda_\tau(\tau), L, \lambda_\alpha(\tau) \} \quad (9) \)

with \( \lambda_\tau(\tau) \neq 0 \) for some nonzero delays \( \tau \).

First, we choose the observed signals \( x(t) \) as \( x_1 \) and \( x(t-\tau) \) as \( x_2 \). Then the Eigen value problem in (4) becomes

\[
R_\alpha(\tau) R_\alpha^{-1}(0) R_\alpha(\tau) a^{(1)}_i = \lambda_i^2 R_\alpha(0) a^{(1)}_i \\
R_\alpha(\tau) R_\alpha^{-1}(0) R_\alpha(\tau) a^{(2)}_i = \lambda_i^2 R_\alpha(0) a^{(2)}_i 
\]

where \( R_\alpha(\tau) \) and \( R_\alpha(0) \) are the correlation matrices of the mixed signals.

In the context of BSS, the vectors \( a^{(1)}_i \) and \( a^{(2)}_i \) in (5) are the same and denote them by \( w_i = a^{(0)}_i = a^{(2)}_i \). Which is the demixing vector applied to the mixed signals. The separate signals can be obtained

\[
y_i(t) = w_i^T x(t). \quad (11) \]

Equation (10) can be rewritten as

\[
R_\alpha^{-1}(0) R_\alpha(\tau) R_\alpha^{-1}(0) R_\alpha(\tau) w_i = \lambda_i^2 w_i \quad (12) \]

which can be further simplified into

\[
R_\alpha(\tau) w_i = \lambda_i R_\alpha(0) w_i, i = 1, \ldots, m. \quad (13) \]

Joint the sub problems into one, we have

\[
R_\alpha(\tau) W = AW^{-1} \quad (14) \]

where the demixing matrix \( W \) contains the vectors \( \{ w_1, \ldots, w_m \} \), \( A = \text{diag}(\lambda_1, \ldots, \lambda_m) \) is a diagonal matrix.

So the CCA problem for \( x(t) \) and \( x(t-\tau) \) can be simplified into a generalized Eigen value decomposition problem, and can be used for BSS [9].

For MCCA, we choose the vector \( x_1, x_2, \ldots, x_k \) as \( x(t-\tau), x(t-\tau), \ldots, x(t-k\tau) \). Compare the (7) and (8). The MCCA problem for \( x(t-\tau), x(t-\tau), \ldots, x(t-k\tau) \) can be simplified into a joint diagonalization problem. We can find a joint diagonalizer \( W \) of \( \{ R_\alpha[\tau_i-\tau], i, j = 1, \ldots, K, i \neq j \} \) using the joint approximate diagonalization method in [14],[15], which satisfies

\[
WR_\alpha[\tau_i-\tau_j]W^T = D_{i,j} \quad (15) \]

where \( \{ D_{i,j} \} \) is a set of diagonal matrices. The separate signals are computed as \( y(t) = W^T x(t) \). We call this BSS algorithm MCCA.

III. KERNELIZING MCCA METHOD FOR NONLINEAR BSS

We generalize the MCCA linear BSS algorithm described above to the invertible nonlinear setting depicted in Fig. 1 [4]. This underlying mixing process

\[
x(t) = f(s(t)) \]

is more general than the post-nonlinear mixture model proposed in [16] because the nonlinearity \( f \) is not restricted to be component wise.

By projecting the input data \( x(t) \) to some high-dimensional feature space via a nonlinear mapping \( \phi \) and then find the weight matrix \( W \) that maximizes the correlation between separate signals. Obviously, if \( \phi = f^{-1} \) and \( W = A^{-1} \), then the sources are perfectly recovered and \( y(t) = s(t) \) [3].

The basic idea of kernel-based method allows constructing very powerful nonlinear variants of existing linear scalar product based algorithms by mapping the data \( x(t) (t = 1, L T) \) implicitly into some kernel feature space \( F \) through some mapping \( \phi : \Re^d \rightarrow \Re^m \). Performing a simple linear algorithm in \( F \), then corresponds to a nonlinear algorithm in input space. All can be done efficiently and never directly but implicitly in \( F \) by using the famous kernel trick \( K_{ij} = \langle \phi(x_i), \phi(x_j) \rangle \). Kernels offer a great deal of flexibility, as they can be generated from other kernels. In the kernel, the data appears only through entries in the Gram matrix. Therefore, this approach gives a further advantage as the number of tunable parameters and updating time does not depend on the number of attributes being used.

A. Constructing reduced kernel feature space

However, a straightforward application of the kernel trick to BSS has so far failed as, after kernelizing, the BSS algorithm has to be applied to a \( T \) dimensional problem which is numerically neither stable nor tractable. We need to specify how to handle its possibly high dimensionality. As in Fig. 2, two methods that obtain an orthogonal basis in feature space with reduced dimension are described [3].

A method to obtain the low dimensional subspace is clustering [4]. Denoting the mapped points by \( \varphi_x := [\varphi(x(1)), L, \varphi(x(T))] \) and \( \varphi_x := [\varphi(v_1), L, \varphi(v_v)] \). We assume that the columns of \( \varphi \) constitute a basis of the column space of \( \varphi \), formally expressed as

\[
\text{span}(\varphi) = \text{span}(\varphi_x) \quad \text{and} \quad \text{rank}(\varphi) = d \quad (16) \]

And \( \varphi \) being a basis implies that the matrix \( \varphi^T \varphi \) has full rank and its inverse exists, an orthonormal basis can be defined as \( \Xi := \varphi \varphi^T \varphi \). This basis enables us to
parameterize the signals in feature space as real valued d-dimensional signals.

\[
\text{input space} \rightarrow \mathbb{R}^m \rightarrow \text{feature space} \rightarrow \mathbb{R}^d \rightarrow \text{reduced feature space}
\]

Figure 2. Mapping the input data to reduced feature space.

\[
\psi_s(t) := \Xi \phi(x(t)) = (\varphi_t, \varphi_r)^{1/2} \varphi_t \phi(x(t)) \quad (17)
\]

By employing the kernel trick

\[
(\varphi_t, \varphi_r)_0 = \phi(v)^T \phi(v) = K(v, v) \quad (18)
\]

where \( i, j = 1, L, d \), \( t = 1, L, T \). Substituting (18) into (17) changes it to

\[
\psi_s(t) := \begin{bmatrix} K(v_i, v_i) & L & K(v_i, v_j) \\ M & O & M \\ K(v_j, v_i) & L & K(v_j, v_j) \end{bmatrix}^{1/2} \begin{bmatrix} K(v_i, x(t)) \\ M \\ K(v_j, x(t)) \end{bmatrix} \quad (19)
\]

Another more direct method to obtain the low dimensional subspace is KPCA (kernel principal component analysis) [4]. For simplicity, we assume that the data is centered in feature space. To perform KPCA we need to find eigenvectors \( E = [e_1, L, e_r] \) and Eigen values \( \lambda_i \geq L \geq \lambda_r \) of the covariance matrix \( \frac{1}{T} \varphi^T \varphi \).

Now, let \( A \) be the diagonal matrix with the Eigen values along the diagonal and let

\[
\left( \frac{1}{T} \varphi_t, \varphi_r \right) E = EA \quad (20)
\]

which can also be expressed as

\[
\left( \frac{1}{T} \varphi_t, \varphi_r \right) (\varphi_s, E) = (\varphi_s, E) A \quad (21)
\]

So \( \lambda_i, L, \lambda_r \) are the Eigen values of \( \frac{1}{T} \varphi_t, \varphi_r \) with corresponding Eigen vectors \( \varphi_s, E \). Normalizing the first \( d \) eigenvectors yields a \( d \)-dimensional orthonormal basis

\[
\Xi := \varphi_s, E \sqrt{(TA)^{-1/2}} \quad (22)
\]

where using \( (TA)^{-1/2} \) to ensure orthonormality and \( E := [e_1, L, e_r] \). So the input data corresponding in the reduced feature space can be described as

\[
\psi_s(t) := \Xi^T \phi(x(t)) = (TA)^{-1/2} E^T \varphi_t \phi(x(t))
\]

\[
= \frac{1}{\sqrt{T}} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} e_1^T \\ M \\ e_r^T \end{bmatrix} \begin{bmatrix} K(x(1), x(t)) \\ M \end{bmatrix} \begin{bmatrix} K(x(T), x(t)) \end{bmatrix} \quad (23)
\]

which are calculated conveniently using the kernel trick.

B. Nonlinear BSS

By mapping the signals \( x(t) \) from input space onto signals \( \psi_s(t) \) in reduced kernel feature space, the nonlinear problem in input space has been transformed to a linear problem in reduced kernel feature space. So we can apply MCCA method to solve this linear BSS problem in reduced kernel feature space.

For the signals in reduced kernel feature space

\[
\psi_s(t) := \Xi^T \phi(x(t)) \in \mathbb{R}^d \quad (24)
\]

We choose the vectors \( \psi_s(t-r_1), \psi_s(t-r_2), K, \psi_s(t-r_k) \) as the components \( x_1, x_2, \ldots, x_k \) in MCCA. The MCCA problem for \( \psi_s(t-r_1), \psi_s(t-r_2), K, \psi_s(t-r_k) \) can be simplified into a joint diagonalization problem.

We can find a joint diagonalizer \( W \) of

\[
\left\{ R_{xx}[r_i-r_j], i, j = 1, \ldots, K, i \neq j \right\}
\]

using the joint approximate diagonalization method in [14],[15], which satisfies

\[
WR_{xx}[r_i-r_j]W^T = \Sigma_{i,j} \quad (25)
\]

where \( \{ \Sigma_{i,j} \} \) is a set of diagonal matrices. The extracted nonlinear components are \( y(t) = W^T \psi_s(t) \), among which are the components of interest. Harmeling and Ziehe, et al. [4] have defined a selection procedure to find the original sources automatically. By repeating the algorithm with the same parameters (kernel choice, \( d \)), but instead of sending \( x(t) \) into the feature space we start with the \( d \) dimensional demixed results \( y(t) \), map them to the feature space, reduce the dimensionality and demix with MCCA, which yield \( y(t) \). The sought-after components of \( y(t) \) are the ones that are matched best by the components \( y(t) \) of the second run of MCCA in feature space.
IV. COMPUTER SIMULATIONS

All simulations were conducted in the MATLAB environment running on an ordinary PC with double 2.2GHZ CPU and 1GB size of memory. To make the demonstration more convincing, the blind separation algorithm was tested on mixtures of realistic speech signals (super-Gaussian signals) and modulated signals (sub-Gaussian signals).

A. Simulations for linear BSS by MCCA

In order to measure the performance of algorithms, we use the performance index (PI) as in [17] defined by

$$PI = \frac{1}{n(n-1)} \sum_{i=1}^{n} \left( \sum_{j=1}^{n} \left| g_{ij} \right| - 1 \right)$$

where $g_{ij}$ is the $(i,j)$-element of the global system matrix $G = HW$ and $\max_i g_{ij}$ represents the maximum value among the elements in the $i$th row vector of $G$, $\max_j g_{ij}$ does the maximum value among the elements in the $i$th column vector of $G$. When the perfect separation is achieved, the performance index is zero. In practice, the values of performance index around $10^{-2}$ gives quite a good performance.

As shown in Fig.3, the source signals are two speech source signals and two modulated signals (a AM signal and a FM signal), which are mixed by random matrix $H$, the sampling points are 20000. The mixing signals are shown as in Fig.4. And the separation results by MCCA BSS algorithm are shown as in Fig.5, the cost time for separation is 0.1563s. It can effectively complete the separate processing and have a good performance of real time.
A lot of experiments have been conducted by changing the impact of additive white Gaussian noise. Results as in Fig. 4 show typical performance of the MCCA BSS algorithm is more accuracy and stabilization than CCA BSS algorithm. When the signal noise ratio (SNR) above zero, the performance index of MCCA BSS algorithm is less than $10^{-3}$, and gives quite a good performance.

### B. Simulations for nonlinear BSS by kernel MCCA

In order to quantificationally evaluate the separation effect, here we choose the resemble coefficient $\xi_{ij}$ of separation signal and source signal to test the resemble degree of separating signal $y(t)$ and expecting signal $s(t)$. The resemble coefficient $\xi_{ij}$ can be described as

$$\xi_{ij} = \frac{\sum_{t=1}^{n} y_i(t) s_j(t)}{\sqrt{\sum_{t=1}^{n} y_i^2(t) \sum_{t=1}^{n} s_j^2(t)}}$$  \hspace{1cm} (27)

where $0 \leq \xi_{ij} \leq 1$. When $y_i(t) = c \cdot s_j(t)$ ($c$ is a constant), then $\xi_{ij} = 1$. It allows separate results exit difference on amplitude. When $y_i(t)$ and $s_j(t)$ are mutual independent, $\xi_{ij} = 0$. When $\xi_{ij} < 0$, it means separate signals and source signals are reverse order.

First simulation is to sub-Gaussian signals. As shown in Fig. 7, the source signals are one sinusoidal signal and one FM signal:

$$x_i(t) = \cos(2\pi \times 1222 t / f_s)$$  \hspace{1cm} (28)

where $t = 1000, f_s = 2000$. These source signals are nonlinearly mixed by

$$x_1(t) = e^{x_1(t)} - e^{-x_1(t)}$$
$$x_2(t) = e^{-x_1(t)} + e^{-x_1(t)}$$ \hspace{1cm} (29)

Employing a Gaussian RBF kernel,

$$K(x_i, x_j) = \exp(-\|x_i - x_j\|^2/2\sigma^2)$$ \hspace{1cm} (30)

and set $d = 20$, using Clustering method to construct the reduced feature space. The mixing signals are shown as in Fig. 8. And the separation results by this proposed kernel MCCA nonlinear BSS algorithm are shown as in
Fig. 9. The resemble coefficients of separate signals and source signals are [0.9913 0.0142; 0.0117 0.9935]. It can effectively complete the separate processing and have a good performance.

The second simulation is to super-Gaussian signals. As shown in Fig. 10, the source signals are two speech source signals (a man and a woman) $s_1(t)=dr1_{_1} fetbo_{_sa2}$ and $s_2(t)=dr2_{_mcewo}_{_sx182}$, were randomly taken from the TIMIT database [3]. These source signals are nonlinearly mixed by

$$
x_1(t) = - (s_1(t) + 1) \cos(\pi s_1(t))
$$
$$
x_2(t) = 1.5(s_2(t) + 1) \sin(\pi s_2(t))
$$

Employing a Gaussian RBF kernel, and set $d=20$, using KPCA method to construct the reduced feature space. The separation results by this proposed kernel MCCA nonlinear BSS algorithm are shown as in Fig. 12. The resemble coefficient matrix of separate signals and source signals is [0.9736 0.0342; 0.0136 0.9927]. It can complete the separate processing successfully.

V. CONCLUSION

This paper proposed an algorithm for nonlinear BSS. It extends the MCCA linear BSS algorithm to the nonlinear domain using the “kernel trick”. In the first step of the method, the input data are mapped to high-dimensional feature space by kernel trick and the effective reduced feature space is extracted by Clustering or KPCA method, translating the nonlinear problem in the input space to a linear problem in reduced feature space. In the second step, by using MCCA linear BSS algorithm, the linear BSS problem can be efficient solved. MCCA linear BSS relies only on second-order statistics of the observation signals, the computation burden is relatively light, can complete the blind separation of super-Gaussian and sub-Gaussian signals at the same time, which make it have a broad application foreground. The experimental results show that the kernel MCCA can complete the nonlinear separate processing successfully.

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REFERENCES

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